Medium Power Amplifier 71 - 86 GHz

Features

- 4 Stage Driver Amplifier for E Band
- 18 dB Gain
- 10 dB Input and Output Match
- 24 dBm Saturated Output Power
- 27 dBm OIP3
- Variable Gain with Adjustable Bias
- Integrated Detector
- Bare die
- RoHS* Compliant
- HBM ESD rating of 200 V
- Size: 3780 x 1500 x 50 µm

Description

The MAAM-011167 is a bare die power amplifier that operates from 71 - 86 GHz. The amplifier provides 18 dB small signal gain. The input and output are matched to 50 Ω with bond wires to external board.

It is designed for use as a driver stage in transmit chains and is ideally suited for E band point to point radios.

Each device is 100% RF tested to ensure performance compliance. The part is fabricated using an efficient pHEMT process.

Ordering Information

Part Number	Package	
MAAM-011167-DIE	Die in Vacuum release ge pack	

Chip Device Layout



Pad Configuration

Pad No.	Function	Pad No.	Function	
1	V _D 1	9	V_{REF}	
2	V _D 2	10	GND _{DET}	
3	V _D 3	11	V _G 4	
4	V _D 4	12	V _G 3	
5	GND	13	V _G 2	
6	RFout	14	V _G 1	
7	GND	15	GND	
8	V _{DET}	16	RF _{IN}	
		17	GND	

* Restrictions on Hazardous Substances, European Union Directive 2011/65/EU.

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Electrical Specifications^{1,2}: Freq. = 71 - 86 GHz, $V_D = 4 V$, $I_D = 360 \text{ mA}$, $T_A = 25^{\circ}C$

Parameter	Units	Min.	Тур.	Max.
Gain	dB	16	18	-
Input Return Loss	dB	-	10	-
Output Return Loss	dB	-	10	-
P1dB	dB	-	21	-
P_{OUT} with P_{IN} = 10 dBm	dBm	20	24	-
P _{SAT} (P3dB)	dBm	-	24	-
OIP3 (worst tone)	dBm	-	27	-
IIP3 (worst tone) for Gain = 20 turned-down to -5 dB	dBm	-	10	-

1. Minimum limits are the on-wafer minimum test limits.

2. Quiescent DC Bias: $I_D 1= 30$ mA, $I_D 2= 60$ mA, $I_D 3= 120$ mA, $I_D 4= 150$ mA. Total DC Power = 1.44 W.

Absolute Maximum Ratings ^{3,4,5}

Parameter	Absolute Max.		
Drain Voltage	4.3 V		
Drain Current	460 mA		
Gate Bias Voltage (V _G 1,2,3,4)	-1.5 V < V _G < +0.3 V		
Input Power	13 dBm		
Storage Temperature	-55°C to +150°C		
Operating Temperature	-40°C to +85°C		
Junction Temperature	150°C		
Thermal Resistance	16.15°C/W		

3. Exceeding any one or combination of these limits may cause permanent damage to this device.

 MACOM does not recommend sustained operation near these survivability limits.

5. Operating at nominal conditions with $T_{\rm J}$ \leq 150°C will ensure MTTF > 1 x 10^6 hours.

Handling Procedures

Please observe the following precautions to avoid damage:

Static Sensitivity

Gallium Arsenide Integrated Circuits are sensitive to electrostatic discharge (ESD) and can be damaged by static electricity. Proper ESD control techniques should be used when handling these class 0 static sensitive devices.

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Gain @ VD = 4 V, I_{DQ} = 360 mA





Output Return Loss @ VD = 4 V, I_{DQ} = 360 mA



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Typical Performance Curves



Input Return Loss @ VD = 4 V, I_{DQ} = 360 mA



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Typical Performance Curves



Input Return Loss @ VD = 4 V, Frequency = 80 GHz



Reverse Isolation @ VD = 4 V, Frequency = 80 GHz



Output Return Loss @ VD = 4 V, Frequency = 80 GHz



Detector Delta Voltage @ VD = 4 V



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Typical Performance Curves

P1dB vs. Frequency @ VD = 4 V, I_{DQ} = 360 mA



Output IP3 vs. Frequency @ VD = 4 V, $I_{DQ} = 250 mA$







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P3dB vs. Frequency @ VD = 4 V, I_{DQ} = 360 mA



Output IP3 vs. Frequency @ VD = 4 V, I_{DQ} = 360 mA



 P_{OUT} , $P_{IN} = 10 \text{ dBm} @ \text{VD} = 4 \text{ V}$, $I_{DQ} = 360 \text{ mA}$



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Typical Performance Curves

Lower Tone Gain vs. Total Current @ VD = 4 V



Lower Tone Input IP3 vs. Total Current @ VD = 4 V



Lower Tone Output IP3 vs. Total Current @ VD = 4 V





Upper Tone Gain vs. Total Current @ VD = 4 V

40 30 20 Gain (dBm) 10 0 -10 81 GHz_Hight 81 GHz_Hight === 86 GHz -20 0 100 200 300 400 Total Drain Current (mA)

Upper Tone Input IP3 vs. Total Current @ VD = 4 V



Upper Tone Output IP3 vs. Total Current @ VD = 4 V



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Calibration Plane

All data was measured on die with 200 μ m pitch probes. The calibration plane is at the middle of the through, 178.5 μ m from the middle of the RF pad.



App Note [1] Biasing -

All gates should be pinched-off ($V_G < -1 V$) before applying drain voltage ($V_D = 4 V$). Then the gate voltages can be increased until the desired quiescent drain current is reached in each stage. The recommended quiescent bias is $V_D = 4 V$, $I_D 1 =$ 30 mA, $I_D 2 = 60$ mA, $I_D 3 = 120$ mA and $I_D 4 = 150$ mA. The performance in this datasheet has been measured with the gate bias set to the voltage that gives the stated value of the quiescent current. It is also possible to regulate the drain current dynamically, to limit the DC power dissipation under RF drive. To turn off the device, the turn on bias sequence should be followed in reverse.

App Note [2] Bias Arrangement -

Each DC pin (V_D1,2,3,4 and V_G1,2,3,4) needs to have bypass capacitance (120 pF and 10 nF) mounted as close to the MMIC as possible.

App Note [3] Wire Bonding -

The loop height of the RF bonds should be minimized. Where the die is mounted above the PCB, it is recommended to use Reverse Ball-Stitch-on-Ball bonds (BSOB). If the die is mounted inside a cavity on the board, Forward Loop bonding may result in a lower loop height.

V-shape RF bond with two wires (diameter = $25 \ \mu m$) is recommended for optimum RF performance.

RF bond wire length to be minimized to reduce the inductance effect. Simulations suggest no more than $300 \mu m$. Substrate RF pad can be optimized to improve the Microstrip to MMIC bond transition as shown in the example below.



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App Note [4] Detector biasing schematic -

As shown in the schematic below, the power detector is biased by matched 120 k Ω resistors to a 5 V bias. The difference voltage between V_{DET} and V_{REF} pins can be obtained using the op-amp differencing circuit shown below.



Layout Dimensions



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Assembly Diagram



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